# Lag-Lead Compensation

Chapter V Control Systems Design by Root-Locus Method

#### Lag—lead Compensation Techniques Based on the Root-Locus Approach.

$$G_c(s) = K_c \left( \frac{s + \frac{1}{T_1}}{s + \frac{\gamma}{T_1}} \right) \left( \frac{s + \frac{1}{T_2}}{s + \frac{1}{\beta T_2}} \right), \qquad (\gamma > 1, \beta > 1)$$

<u>Case I  $\gamma \neq \beta$ .</u> In this case, the design process is a combination of the design of the lead compensator and that of the lag compensator. The design procedure for the lag-lead compensator follows:

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- 1. From the given performance specifications, determine the
- There are a specifications, determine the desired location for the dominant closed-loop poles.
   Using the uncompensated open-loop transfer function G(s), determine the angle deficiency φ if the dominant closed-loop poles are to be at the desired location. The the desired location is a specification of the dominant closed does a specification of phase-lead portion of the lag-lead compensator must
- contribute this angle Ø.
  3. Assuming that we later choose T₂ sufficiently large so that the magnitude of the lag portion \_\_\_\_\_\_



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is approximately unity, where  $s = s_1$  is one of the dominant closed-loop poles, choose the values of  $T_1$  and  $\gamma$  from the requirement that

$$\frac{s_1 + \frac{1}{T_1}}{s_1 + \frac{\gamma}{T_1}} = \phi$$

The choice of  $T_1$  and  $\gamma$  is not unique. (Infinitely many sets of  $T_1$  and  $\gamma$  are possible.) Then determine the value of  $K_c$  from the magnitude condition:

$$\left| \frac{s_1 + \frac{1}{T_1}}{s_1 + \frac{\gamma}{T_1}} G(s_1) \right| = 1$$

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4. If the static velocity error constant K<sub>w</sub> is specified, determine the value of β to satisfy the requirement for K<sub>w</sub>. The static velocity error constant K<sub>w</sub> is given by
K<sub>w</sub> = lim<sub>x</sub> sG<sub>c</sub>(s)G(s)

$$= \lim_{s \to 0} s \mathcal{K}_{\epsilon} \left( \frac{s + \frac{1}{T_1}}{s + \frac{\gamma}{T_1}} \right) \left( \frac{s + \frac{1}{T_2}}{s + \frac{1}{\beta T_2}} \right) G(s)$$
$$= \lim_{s \to 0} s \mathcal{K}_{\epsilon} \frac{\beta}{\gamma} G(s)$$

where  $K_c$  and  $\gamma$  are already determined in step 3. Hence, given the value of  $K_v$  the value of  $\beta$  can be determined from this last equation. Then, using the value of  $\beta$  thus determined,

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Choose the value of  $T_2$  such that

$$\begin{vmatrix} \frac{s_{1} + \frac{1}{T_{2}}}{s_{1} + \frac{1}{\beta T_{2}}} \end{vmatrix} \doteq 1$$
$$-5^{\circ} < \underbrace{\int \frac{s_{1} + \frac{1}{T_{2}}}{s_{1} + \frac{1}{T_{2}}} < 0^{\circ}}_{\frac{s_{1} + \frac{1}{\beta T_{2}}}{s_{1} + \frac{1}{\beta T_{2}}}} < 0^{\circ}$$

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Case 2,  $\gamma = \beta$ . If  $\gamma = \beta$  is required in Equation (7-6), then the preceding design recedure for the lag-lead compensator may be modified as follows: **1.** From the given performance specifications, determine the desired location for the dominant closed-loop poles.

2. The lag-lead compensator given by Equation (7-6) is modified to

$$G_{c}(s) = K_{c} \frac{(T_{1}s+1)(T_{2}s+1)}{\left(\frac{T_{1}}{\beta}s+1\right)(\beta T_{2}s+1)} = K_{c} \frac{\left(s+\frac{1}{T_{1}}\right)\left(s+\frac{1}{T_{2}}\right)}{\left(s+\frac{1}{\beta}\right)\left(s+\frac{1}{\beta T_{2}}\right)}$$

(7–7)

where  $\beta > 1$ . The open-loop transfer function of the compensated system is  $G_{c}(s)G(s)$ . If the static velocity error constant  $K_{c}$  is specified, determine the value of constant  $K_{c}$  from the following equation:

$$K_{p} = \lim_{s \to 0} sG_{c}(s)G(s)$$
$$= \lim_{s \to 0} sK_{c}G(s)$$

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3. To have the dominant closed-loop poles at the desired location, calculate the angle  
contribution 
$$\phi$$
 needed from the phase lead portion of the lag-lead compensator.  
4. For the lag-lead compensator, we later choose  $T_2$  sufficiently large so that  

$$\begin{aligned} & \left| \frac{s_1 + \frac{1}{T_1}}{s_1 + \frac{1}{BT_2}} \right| \\
\text{is approximately unity, where } s = s_1 \text{ is one of the dominant closed-loop poles. Determine the values of  $T_1$  and  $\beta$  from the magnitude and angle conditions:  

$$\begin{aligned} & \left| K_n \left( \frac{s_1 + \frac{1}{T_1}}{s_1 + \frac{\beta}{T_1}} \right) G(s_1) \right| = 1 \\
& \left| \frac{s_1 + \frac{1}{T_1}}{s_1 + \frac{\beta}{T_1}} \right| = \phi \end{aligned}$$$$

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5. Using the value of  $\beta$  just determined, choose  $T_2$  so that

$$\begin{vmatrix} s_1 - \frac{1}{T_2} \\ \frac{s_1 - 1}{\beta T_2} \end{vmatrix} \approx 1$$
  
- $S^* < \int \frac{s_1 + 1}{s_1 - \frac{1}{\beta T_2}} < 0^{\circ}$ 

The value of  $\beta T_i$ , the largest time constant of the lag-lead compensator, should not be too large to be physically realized. (An example of the design of the lag-lead compensator when  $\gamma = \beta$  is given in Example 7–4.)









Since the requirement on the static velocity error constant is 80 sec  $^{\rm 1},$  We have

Example II: Lag-Lead Compensator Case:  $\gamma = \beta$ For the phase lag portion, we may choose Thus, the lag-lead compensator becomes

## Example II: Lag-Lead Compensator Case: $\gamma = \beta$

The compensated system will have the open-loop transfer function

The closed-loop transfer function

New closed-loop poles and zeros are located at





